

LQG Robust Stability

MEM 355 Performance Enhancement of Dynamical Systems

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Outline

- Robust Stability
 - Min-Max Control
 - Classical LQG
 - Stability Margins & LTR

Regulation: Steer the system from any initial state to the origin.

Min-Max Control

$$\dot{x} = Ax + Bu + Ew$$

$$y = Cx$$

$$J(u, w) = \int_0^{\infty} \left[y^T(t)y(t) + \rho u^T(t)u(t) - \gamma^2 w^T(t)w(t) \right] dt, \quad \rho, \gamma > 0$$

objective: $\min_u \max_w J(u, w)$

Theorem: Suppose $w(t)$ has bounded energy, $(A, B), (A, E)$ stabilizable, (A, C) detectable, then

the optimal control is $u^*(t) = Kx(t)$, $K = -\frac{1}{\rho} B^T S$

the worst case disturbance is $w^*(t) = \frac{1}{\gamma^2} E^T Sx(t)$

S is the unique, symmetric, nonnegative solution of the Riccati equation:

$$A^T S + SA - S \left(\frac{1}{\rho} BB^T - \frac{1}{\gamma^2} EE^T \right) S = -C^T C$$

Min-Max Hamiltonian Matrix

$$A^T S + SA - S \left(\frac{1}{\rho} BB^T - \frac{1}{\gamma^2} EE^T \right) S = -C^T C \Rightarrow H = \begin{bmatrix} A & \frac{1}{\gamma^2} EE^T - \frac{1}{\rho} BB^T \\ -C^T C & -A^T \end{bmatrix}$$

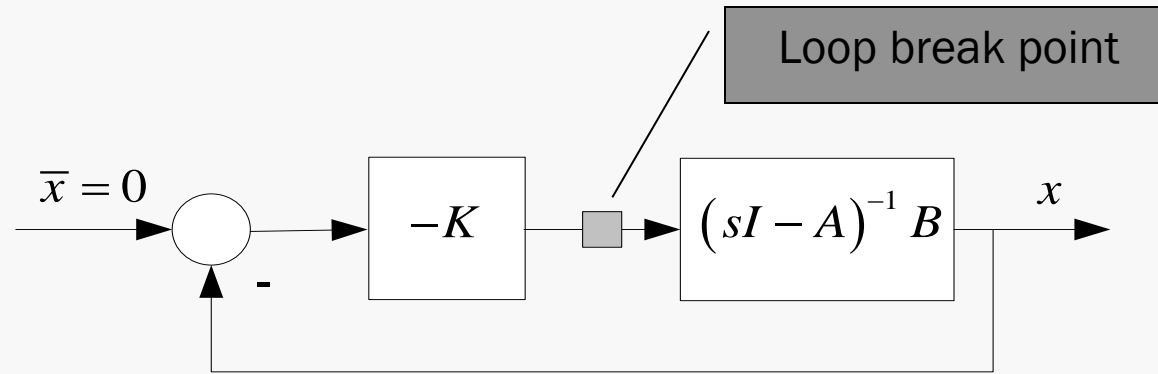
stabilizability/detectability \Rightarrow

- $\exists \gamma_{\min}$ such that there are no pure imaginary eigenvalues of H if $\gamma > \gamma_{\min}$
- $\gamma \rightarrow \infty$ produces LQR solution
- $\gamma = \gamma_{\min}$ is the full state feedback H_{∞} solution
- all $\gamma_{\min} \leq \gamma < \infty$ are valid min-max controllers

Min-Max, Continued

- $\operatorname{Re} \lambda(H) \neq 0 \Leftrightarrow \operatorname{Re} \lambda \left(A + \frac{1}{\gamma^2} EE^T S - \frac{1}{\rho} BB^T S \right) < 0$
- The closed loop system matrix is $A - \frac{1}{\rho} BB^T S$, since $\frac{1}{\gamma^2} EE^T S$ is destabilizing, there is some induced stability margin

LQG Robustness: State Feedback



$$G(s) = -K (sI - A)^{-1} B = -K \Phi B, \quad \Phi = (sI - A)^{-1}$$

$$S(s) = \left[I - K (sI - A)^{-1} B \right]^{-1}$$

For R diagonal, $\bar{\sigma}(S(j\omega)) \leq 1 \forall \omega \Rightarrow$

$$GM : \frac{1}{2}, \infty \quad PM : \frac{\pi}{3} \quad (\text{in each channel})$$

LQG Robustness: State Feedback 2

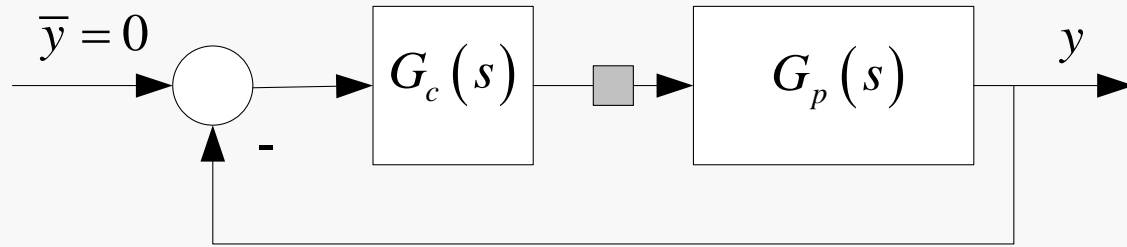
When K is determined via the Riccati equation, it is known

$$\sigma_{\min} [S^{-1}] = \min_{\|v\|_R=1} \left\{ \left\| \left[I - K (j\omega I - A)^{-1} B \right] v \right\|_R \right\} \geq 1, \omega > 0$$

Similarly,

$$\min_{\|v\|_V=1} \left\{ \left\| \left[I - C (j\omega I - A)^{-1} L \right] v \right\|_V \right\} \geq 1, \omega > 0$$

LQG Robustness: Output Feedback

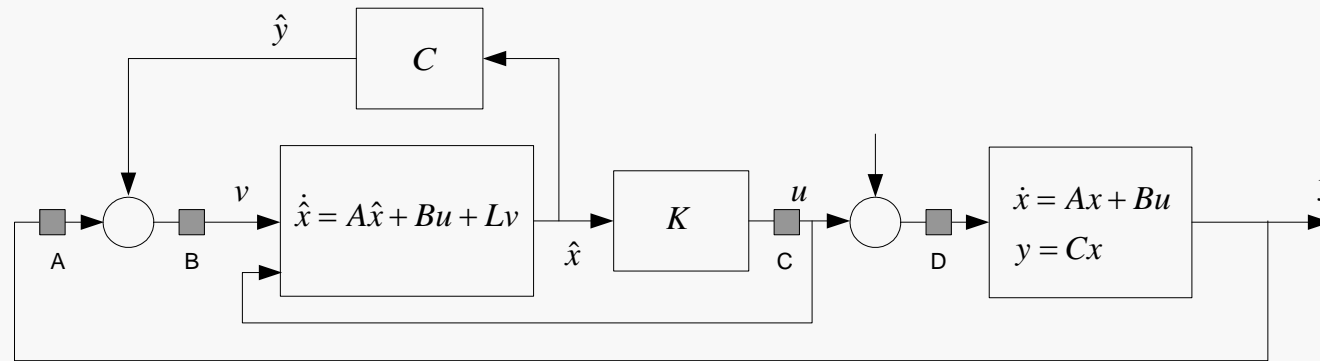


$$G_p(s) = C(sI - A)^{-1} B$$

$$G_c(s) = -K \left[s - (A + BK + LC) \right]^{-1} L$$

No guaranteed margins!

Output Feedback-a closer look



$$\Phi(s) \triangleq [sI - A]^{-1}$$

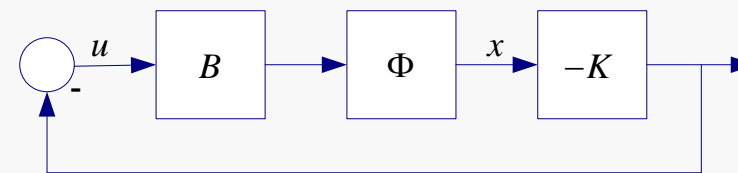
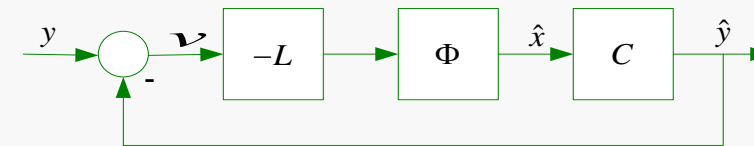
Loop transfer functions:

(A) $C\Phi BK [\Phi^{-1} - BK - LC]^{-1} L$

(B) $-C\Phi L$

(C) $-K\Phi B$

(D) $K [\Phi^{-1} - BK - LC]^{-1} LC\Phi B$



Loop breaking at (A) & (D) do not produce the excellent margins of state feedback.

Example

$$\dot{x}(t) = Ax(t) + Bu(t) + Ed(t)$$

$$y(t) = Cx(t) + 0.1v(t)$$

$$x = \begin{bmatrix} \theta \\ \gamma \\ q \\ V \\ x_w \end{bmatrix} \begin{array}{l} \text{pitch angle} \\ \text{flight path angle, } \gamma = \theta - \alpha \\ \text{pitch rate} \\ \text{speed} \\ \text{wind gust state} \end{array} \quad u = \begin{bmatrix} \delta_e \\ \delta_f \end{bmatrix} \begin{array}{l} \text{elevator} \\ \text{flaperon} \end{array}$$

$$A = \begin{bmatrix} 0.0 & 0.0 & 1.0 & 0.0 & 0.0 \\ 1.50 & -1.50 & 0.0 & 0.0057 & 1.50 \\ -12.0 & 12.0 & -0.60 & -0.0344 & -12.0 \\ -0.8520 & 0.290 & 0.0 & -0.0140 & -0.290 \\ 0.0 & 0.0 & 0.0 & 0.0 & -0.730 \end{bmatrix}$$

$$B = \begin{bmatrix} 0.0 & 0.0 \\ 0.160 & 0.80 \\ -19.0 & -3.0 \\ -0.0115 & -0.0087 \\ 0.0 & 0.0 \end{bmatrix} \quad L = \begin{bmatrix} 0.0 \\ 0.0 \\ 0.0 \\ 0.0 \\ 1.1459 \end{bmatrix} \quad C = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 & 0 \end{bmatrix}$$

Loop Transfer Recovery

Recall $L = SC^T V^{-1}$, with

$$SA^T + AS - SC^T V^{-1} CS = -W$$

Let $W = W_0 + \rho^2 BB^T$, ρ a scalar. It can be shown, for a minimum phase plant

$$\lim_{\rho \rightarrow \infty} G_c(s) G_p(s) = K(sI - A)^{-1} B \quad \text{complete recovery!}$$

Correspondingly,

$$\lim_{\rho \rightarrow \infty} G_c(s) = K(sI - A)^{-1} B G_p^{-1}(s)$$

\Rightarrow some of the estimator poles tend to the plant zeros

For nonminimum phase plants, estimator poles tend to LHP plant zeros, recovery is not complete.